

Q1)

(a)

(i) Assuming that all acceptors are ionized

$$N_A = p = N_V e^{\left[\frac{-E_F}{k_B T}\right]} = 8 \times 10^{24} \exp[-0.06/(8.617 \times 10^{-5} \times 298)] = 7.7 \times 10^{23} \text{ m}^{-3}$$

(ii) From the Law of Mass Action:

$$n_i = \sqrt{N_C N_V} e^{\left(\frac{-E_g}{2k_B T}\right)}$$

Hence,

$$N_C = n_i^2 / N_V e^{\left(\frac{E_g}{k_B T}\right)} = 10^{32} / (8 \times 10^{24}) \exp[1.1/(8.617 \times 10^{-5} \times 298)] = 5.02 \times 10^{25} \text{ m}^{-3}$$

From this, the density of donors can be found, again assuming that all are ionized:

$$N_D = n$$

$$E_F - E_C = k_B T \ln\left(\frac{N_D}{N_C}\right) = 8.617 \times 10^{-5} \times 298 \times \ln\left(\frac{7.7 \times 10^{23}}{5.02 \times 10^{25}}\right) = -0.107 \text{ eV}$$

(b) (i) Consider the p-type side of the junction. The Poisson equation states that

$$\nabla^2 V = \frac{-\rho}{\epsilon_0 \epsilon_r} = \frac{-eN_A}{\epsilon_0 \epsilon_r}$$

Given that V only varies in the x direction, then

$$\frac{d^2 V}{dx^2} = \frac{-eN_A}{\epsilon_0 \epsilon_r}$$

Integrating the equation gives the general solution,

$$\frac{dV}{dx} = \frac{-eN_A x}{\epsilon_0 \epsilon_r} + C$$

We know that at $x = w_p$ the electric field is zero, so

$$C = \frac{eN_A w_p}{\epsilon_0 \epsilon_r}$$

and hence,

$$\frac{dV}{dx} = \frac{eN_A (w_p - x)}{\epsilon_0 \epsilon_r}$$

Integrating again gives the general solution,

$$V = \frac{eN_A}{\epsilon_0 \epsilon_r} \left(K + w_p x - \frac{x^2}{2} \right)$$

We can make $x = 0$ our arbitrary reference point, and set V at this point to be zero, so $K = 0$ and then the voltage on the p-side, V_p is simply the potential at $x = w_p$, so

$$V_p = \frac{eN_A w_p^2}{2\epsilon_0 \epsilon_r}$$

As the n-type side of the junction has an equal doping density, then the width of the junction on the n side must be the same, as must be the potential difference, so

$$V_p = \frac{eN_A w_p^2}{2\epsilon_0 \epsilon_r} = V_n = \frac{eN_D w_n^2}{2\epsilon_0 \epsilon_r} = \frac{V_0}{2}$$

Hence,

$$w = 2w_p = 2 \left(\frac{\epsilon_0 \epsilon_r V_0}{eN_A} \right)^{1/2}$$

b(ii) For a pn junction

$$eV_0 = E_{Fn} - E_{Fp} = E_C + (E_{fn} - E_C) - E_{Fp} = 1.1 + (-0.107) - 0.06 = 0.933 \text{ eV}$$

Hence

$$w = 2 \left(\frac{12 \times 8.854 \times 10^{-12} \times 0.933}{1.602 \times 10^{-19} \times 7.7 \times 10^{23}} \right)^{0.5} = 56.7 \text{ nm}$$

Q2)

(a)

$$N = \int_0^\infty g(E) f(E) dE = \int_0^{E_F} g(E) dE$$

Substituting $g(E)dE = \frac{V}{2\pi^2\hbar^3} (2m)^{\frac{3}{2}} E^{\frac{1}{2}} dE$ we get

$$N = \frac{V}{2\pi^2\hbar^3} (2m)^{\frac{3}{2}} \int_0^{E_F} E^{\frac{1}{2}} dE = \frac{V}{2\pi^2\hbar^3} (2m)^{\frac{3}{2}} \frac{E_F^{\frac{3}{2}}}{\frac{3}{2}}$$

Therefore:

$$E_F = \frac{\hbar^2}{2m} \left(3\pi^2 \frac{N}{V} \right)^{\frac{2}{3}}$$

(b) Each atom contributes 2 free electrons:

$$N/V = 2 \times 3.92 \times 10^{28} = 7.84 \times 10^{28} \text{ m}^{-3}$$

Hence:

$$E_F = \frac{\hbar^2}{2m} \left(3\pi^2 \frac{N}{V} \right)^{\frac{2}{3}} = 6.69 \text{ eV}$$

(c)

$$E = \frac{\hbar^2 |\mathbf{k}|^2}{2m}$$

$$k = \frac{\sqrt{2mE}}{\hbar} = 1.324 \times 10^{10} \text{ m}^{-1}$$

$$\lambda = \frac{2\pi}{k} = 4.75 \times 10^{-10} \text{ m}$$

$$P = \hbar k = mv$$

$$v = \frac{\hbar k}{m} = 1.53 \times 10^6 \text{ m/s}$$

Problem 3a

- (i) $\because \delta n_p(x) > n_{p0}$
 $\Rightarrow$ diode is forward biased
 and $\delta n_n(x) > n_{n0}$

- (ii) The equilibrium minority carrier concentration:

$$n_{p0} < p_{n0}$$

$$\Rightarrow \frac{n_i^2}{p_{n0}} < \frac{n_i^2}{n_{n0}} \Rightarrow \boxed{n_{n0} < p_{p0}}$$

$\hookrightarrow$ majority carrier concentration on the n-side is smaller than on the p-side.

- (iii) Emission of light from a p-n junction corresponds to energy loss in the form of electromagnetic radiation. Energy loss occurs due to recombination of electrons and holes in the junction or depletion region. Hence, the diode has to be forward biased for the carriers to cross the depletion region and recombine to emit light.

(iv) Two ideal diodes at $T=300K$

Diode 1: $E_{g1} = 0.525 \text{ eV}$

$I_1 = 10 \text{ mA}$ at $V_{a1} = 0.255 \text{ V}$

Diode 2: $E_{g2} = ??$

$I_2 = 10 \mu\text{A}$ at $V_{a2} = 0.32 \text{ V}$

The current in a FB diode is given by:

$$I = A \cdot J_s \cdot \left[\exp\left(\frac{eV_a}{kT}\right) - 1 \right]$$

$$\Rightarrow I = A \cdot \left[\frac{eD_p p_{n0}}{L_p} + \frac{eD_n n_{p0}}{L_n} \right] \cdot \left[\exp\left(\frac{eV_a}{kT}\right) - 1 \right]$$

$$\Rightarrow I = A \cdot e \cdot n_i^2 \left[\frac{D_p}{L_p N_D} + \frac{D_n}{L_n N_A} \right] \cdot \left[\exp\left(\frac{eV_a}{kT}\right) - 1 \right] \quad \text{--- (1)}$$

$\because$ the two diodes have exact same physical parameters except bandgap $\Rightarrow$ the only parameter in eqn (1) dependent on E_g is

$$n_i^2 \quad \because n_i^2 \propto \exp\left\{ \frac{-E_g}{kT} \right\}$$

Hence we can write,

$$I_1 \propto n_{i1}^2 \cdot \left[\exp\left\{ \frac{eV_{a1}}{kT} \right\} - 1 \right] \propto \exp\left\{ \frac{-E_{g1}}{kT} \right\} \cdot \left[\exp\left(\frac{eV_{a1}}{kT} \right) - 1 \right]$$

$$\text{and } I_2 \propto n_{i2}^2 \left[\exp\left(\frac{eV_{a2}}{kT} \right) - 1 \right] \propto \exp\left\{ \frac{-E_{g2}}{kT} \right\} \cdot \left[\exp\left(\frac{eV_{a2}}{kT} \right) - 1 \right]$$

$$\therefore \frac{I_1}{I_2} = \exp\left\{ \frac{E_{g2} - E_{g1}}{kT} \right\} \cdot \frac{\left[\exp\left(\frac{eV_{a1}}{kT} \right) - 1 \right]}{\left[\exp\left(\frac{eV_{a2}}{kT} \right) - 1 \right]}$$

$$\Rightarrow \exp\left(\frac{E_{g2} - E_{g1}}{kT} \right) = \frac{I_1}{I_2} \cdot \frac{\left[\exp\left(\frac{eV_{a2}}{kT} \right) - 1 \right]}{\left[\exp\left(\frac{eV_{a1}}{kT} \right) - 1 \right]}$$

$$\Rightarrow (E_{g2} - E_{g1}) = KT \ln \left\{ \frac{I_1}{I_2} \frac{\left[\exp \left\{ \frac{eV_{g2}}{KT} \right\} - 1 \right]}{\left[\exp \left\{ \frac{eV_{g1}}{KT} \right\} - 1 \right]} \right\}$$

$$\Rightarrow E_{g2} - E_{g1} = (0.0259) \ln \left\{ \frac{10 \times 10^{-3}}{10 \times 10^{-6}} \cdot \frac{\exp \left\{ \frac{0.32}{0.0259} \right\} - 1}{\exp \left\{ \frac{0.255}{0.0259} \right\} - 1} \right\}$$

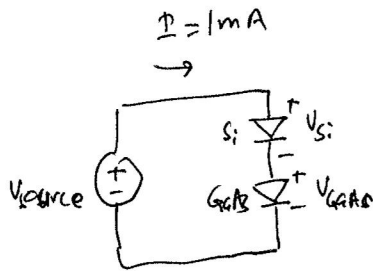
$$\Rightarrow E_{g2} - E_{g1} = 0.2349 \text{ eV}$$

$$\therefore E_{g1} = 0.525 \text{ eV}$$

$$\Rightarrow \boxed{E_{g2} = 0.769 \text{ eV}} \text{ Ans.}$$

Problem 3b

(i)



$$V_{\text{source}} = V_{\text{Si}} + V_{\text{GaAs}}$$

Area of both diodes: $A = 100 \times 100 \mu\text{m}^2 = 10^{-4} \text{cm}^2$

$$N_A = 5 \times 10^{15} / \text{cm}^3$$

$$N_D = 1 \times 10^{18} / \text{cm}^3$$

Circuit current = 1mA

for ~~full~~ from Shockley equation:

$$I = I_s \left\{ \exp\left(\frac{eV_a}{kT}\right) - 1 \right\}$$

$$\Rightarrow I = A \cdot n_i^2 \cdot e \left[\frac{D_p}{L_p N_D} + \frac{D_n}{L_n N_A} \right] \cdot \left[\exp\left\{ \frac{eV_a}{kT} \right\} - 1 \right]$$

$$\Rightarrow I = A \cdot n_i^2 \cdot e \cdot \left[\frac{D_p}{\sqrt{D_p \tau_p} N_D} + \frac{D_n}{\sqrt{D_n \tau_n} N_A} \right] \cdot \left[\exp\left\{ \frac{eV_a}{kT} \right\} - 1 \right]$$

For Si Diode:

$$1 \times 10^{-3} = 10^{-4} \cdot (10^{10})^2 \cdot (1.6 \times 10^{-19}) \left[\frac{12}{\sqrt{12 \times 10^{-6}} \cdot 10^{18}} + \frac{35}{\sqrt{35 \times 10^{-6}} \cdot 5 \times 10^{15}} \right] \cdot \left[\exp\left\{ \frac{eV_a}{kT} \right\} - 1 \right]$$

$$1 \times 10^{-3} = 1.9 \times 10^{-15} \left[\exp\left\{ \frac{eV_a}{kT} \right\} - 1 \right]$$

$$\Rightarrow V_a = \frac{kT}{e} \cdot \ln \left[\frac{10^{-3}}{1.9 \times 10^{-15}} + 1 \right]$$

$$\Rightarrow \underline{V_a = 0.70 \text{V}} \quad [= V_{\text{Si}}]$$

for GaAs Diode:

$$1 \times 10^{-3} = 10^{-4} (2 \times 10^6)^2 (1.6 \times 10^{-19}) \left[\frac{10}{\sqrt{10 \times 10^{-8}} \cdot 10^{18}} + \frac{220}{\sqrt{220 \times 10^{-8}} \cdot 5 \times 10^{15}} \right] \cdot \left[\exp\left\{ \frac{eV_a}{kT} \right\} - 1 \right]$$

$$\Rightarrow 1 \times 10^{-3} = 1.9 \times 10^{-21} \left[\exp\left\{ \frac{eV_a}{kT} \right\} - 1 \right]$$

$$\Rightarrow \underline{V_a = 1.06 \text{V}} \quad [= V_{\text{GaAs}}]$$

Hence: $V_{\text{source}} = V_{\text{Si}} + V_{\text{GaAs}}$

$$\Rightarrow \underline{V_{\text{source}} = 1.76 \text{V}}$$

(ii)

The difference in forward voltages originates from the fact that the threshold voltage (voltage at which the diode turns on) is approximately equal to $\frac{E_g}{e}$. Since Silicon has a smaller bandgap [$E_{gSi} = 1.12 \text{ eV}$] compared to GaAs [$E_{gGaAs} = 1.42 \text{ eV}$] $\Rightarrow$ Silicon has a small turn on voltage.

(iii)

$\therefore$ Bandgap of Ge $E_{gGe} = 0.66 \text{ eV} < E_{gSi} < E_{gGaAs}$

Hence V_a for Ge will be small than both Si and GaAs.

Problem 4a

(i)

MOSFET-1: n-channel or NMOS

MOSFET-2: p-channel or PMOS

(ii) SUBSTRATE: p-type

WELL: n-type

(iii) MOSFET 1:

Metal workfunction (Aluminum) $\phi_m = 4.2 \text{ eV}$

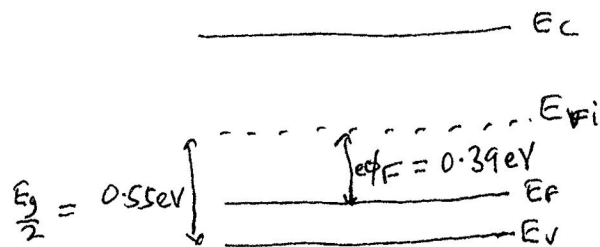
Electron Affinity of Silicon $\chi_s = 4.0 \text{ eV}$

Substrate = p-type, doping $N_A = 5 \times 10^{16} / \text{cm}^3$

$$\phi_F = \frac{kT}{e} \ln \left(\frac{N_A}{n_i} \right) = 0.0259 \ln \left(\frac{5 \times 10^{16}}{1.5 \times 10^{10}} \right) = 0.39 \text{ V}$$

[$\phi_F > 0$ for p-type]

or $E_{Fi} - E_F = 0.39 \text{ eV}$



Hence! Semiconductor work function:

$$\phi_s = \chi_s + \frac{E_g}{2e} + \phi_F = 4.0 + 0.55 + 0.39$$

$$\Rightarrow \phi_s = 4.94 \text{ V}$$

or work function difference

$$\phi_{ms} = \phi_m - \phi_s = 4.2 - 4.94 = \underline{\underline{-0.74 \text{ V}}}$$

Flat band voltage: $V_{FB} = \phi_{ms} = \underline{\underline{-0.74 \text{ V}}}$

(iv)

MOSFET 2:

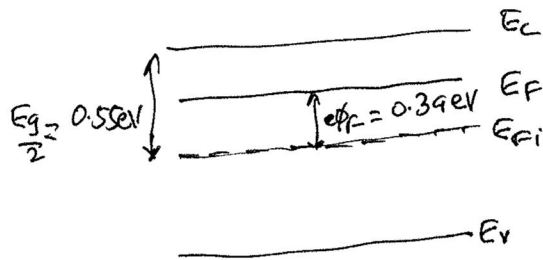
$$\phi_m (\text{Aluminum}) = 4.2 \text{ eV}$$

Substrate well = ~~n~~-type

$$\Rightarrow \text{Doping } N_D = 5 \times 10^{16} / \text{cm}^3$$

$$\phi_F = -\frac{kT}{e} \ln\left(\frac{N_D}{n_i}\right) = -0.39 \text{ V} \quad [\phi_F < 0 \text{ for n-type}]$$

$$\text{or } E_F - E_{Fi} = 0.39 \text{ eV}$$



$\therefore$ Semiconductor work function:

$$\phi_s = \frac{\chi_s}{e} + \frac{E_g}{2e} + \phi_F$$

$$\Rightarrow \phi_s = 4 + 0.55 - 0.39$$

$$\Rightarrow \phi_s = 4.16 \text{ V}$$

work function difference,

$$\phi_{ms} = \phi_m - \phi_s = 4.2 - 4.16 = \underline{\underline{0.04 \text{ V}}}$$

or Flat Band Voltage

$$V_{FB} = \underline{\underline{0.04 \text{ V}}}$$

(v)

MOSFET 1: Threshold voltage:

$$V_{T, \text{ideal}} = 2\phi_F + \frac{E_{si}}{e}$$

$$V_{T, \text{ideal}} = 2\phi_F + \frac{e \cdot t_{ox}}{e \cdot \epsilon_{ox}} \cdot \sqrt{\frac{2q \cdot N_A \cdot 2\phi_F}{\epsilon_{si}}}$$

$$\Rightarrow V_{T, \text{ideal}} = 2(0.4) + \frac{1.7}{3.9} \cdot 500 \times 10^{-8} \cdot \sqrt{\frac{2(1.6 \times 10^{-19})(5 \times 10^{16}) \cdot 2(0.4)}{11.7 \times (8.85 \times 10^{-14})}}$$

$$\Rightarrow V_{T, \text{ideal}} = 2.46 \text{ V}$$

$$V_{T,real} = V_{T,ideal} - |V_{FB}|$$

$$\Rightarrow V_{T,real} = 2.46 - 0.74 = \underline{\underline{1.72V}}$$

MOSFET 2: Threshold voltage:

$$V_{T,ideal} = -(\underbrace{2\phi_F}_{\phi_s}) - \underbrace{\frac{C_{ox}}{C_{ox}}}_{V_{ox}} \cdot \sqrt{\frac{2 \cdot q N_D \cdot 2\phi_F}{\epsilon_s}}$$

[NOTE: Both ϕ_s and V_{ox} are negative]

$$\Rightarrow V_{T,ideal} = -2.46V$$

$$\therefore V_{T,real} = -2.46 + 0.04 = \underline{\underline{-2.42V}}$$

NOTE:

MOSFET 1 $\rightarrow$ NMOS $\rightarrow V_{T,real}$ is positive

MOSFET 2 $\rightarrow$ PMOS $\rightarrow V_{T,real}$ is negative

} correct!!

(i)

Given: $A = 7 \times 10^{-4} \text{ cm}^2$, $T = 300 \text{ K}$, $I_{\text{total}} = 0.8 \text{ mA}$, $J_{\text{total}} = I_{\text{total}}/A$

Schoetky:

$$J_{\text{total}} = J_s \left\{ \exp\left(\frac{eV_a}{kT}\right) - 1 \right\}$$

$$J_s = 4 \times 10^{-8} \frac{\text{A}}{\text{cm}^2}$$

$$\therefore V_a = \frac{kT}{e} \left[\ln\left(\frac{J_{\text{total}}}{J_s}\right) + 1 \right]$$

$$\therefore V_a = 0.0259 \left[\ln\left[\frac{0.8 \times 10^{-3}}{7 \times 10^{-4} \times 4 \times 10^{-8}}\right] + 1 \right]$$

$$\Rightarrow V_a = \underline{\underline{0.4447 \text{ Volts.}}}$$

p-n junction

$$J_{\text{total}} = J_s \left\{ \exp\left(\frac{eV_a}{kT}\right) - 1 \right\}$$

$$J_s = 3 \times 10^{-12} \frac{\text{A}}{\text{cm}^2}$$

$$V_a = \frac{kT}{e} \left\{ \ln\left(\frac{J_{\text{total}}}{J_s}\right) + 1 \right\}$$

$$V_a = 0.0259 \left\{ \ln\left[\frac{0.8 \times 10^{-3}}{7 \times 10^{-4} \times 3 \times 10^{-12}}\right] + 1 \right\}$$

$$V_a = \underline{\underline{0.6987 \text{ Volts}}}$$

(ii)

p-n diode I_{total} at $T = 400 \text{ K}$?

$$\therefore J_s \propto n_i^2 \propto T^3 \exp\left\{-\frac{E_g}{kT}\right\} \Rightarrow \frac{J_s(400 \text{ K})}{J_s(300 \text{ K})} = \left(\frac{400}{300}\right)^3 \cdot \frac{\exp\left\{-\frac{1.12}{0.0259 \times \frac{400}{300}}\right\}}{\exp\left\{-\frac{1.12}{0.0259}\right\}}$$

$$\Rightarrow J_s(400 \text{ K}) = 3.52 \times 10^{-7} \frac{\text{A}}{\text{cm}^2}$$

$$\therefore I_{\text{total}}(400 \text{ K}) = A \cdot J_s(400 \text{ K}) \cdot \exp\left\{\frac{eV_a}{kT_{400 \text{ K}}}\right\} = 7 \times 10^{-4} \times 3.52 \times 10^{-7} \times \exp\left\{\frac{0.6907}{0.0259 \times \frac{400}{300}}\right\}$$

$$\Rightarrow I_{\text{total}}(400 \text{ K}) = \underline{\underline{119 \text{ mA}}}$$

Schoetky: I_{total} at $T = 400 \text{ K}$?

$$J_s \propto T^2 \exp\left\{-\frac{e\phi_b}{kT}\right\} \Rightarrow \frac{J_s(400 \text{ K})}{J_s(300 \text{ K})} = \left(\frac{400}{300}\right)^2 \cdot \frac{\exp\left\{-\frac{0.82}{0.0259 \times \frac{400}{300}}\right\}}{\exp\left\{-\frac{0.82}{0.0259}\right\}}$$

$$\Rightarrow J_s(400 \text{ K}) = 1.95 \times 10^{-4} \frac{\text{A}}{\text{cm}^2}$$

$$\therefore I_{\text{total}}(400 \text{ K}) = A \cdot J_s(400 \text{ K}) \cdot \exp\left\{\frac{eV_a}{kT_{400 \text{ K}}}\right\} = 7 \times 10^{-4} \times 1.95 \times 10^{-4} \times \exp\left\{\frac{0.4447}{0.0259 \left(\frac{400}{300}\right)}\right\}$$

$$\Rightarrow I_{\text{total}} = \underline{\underline{53.42 \text{ mA}}}$$